

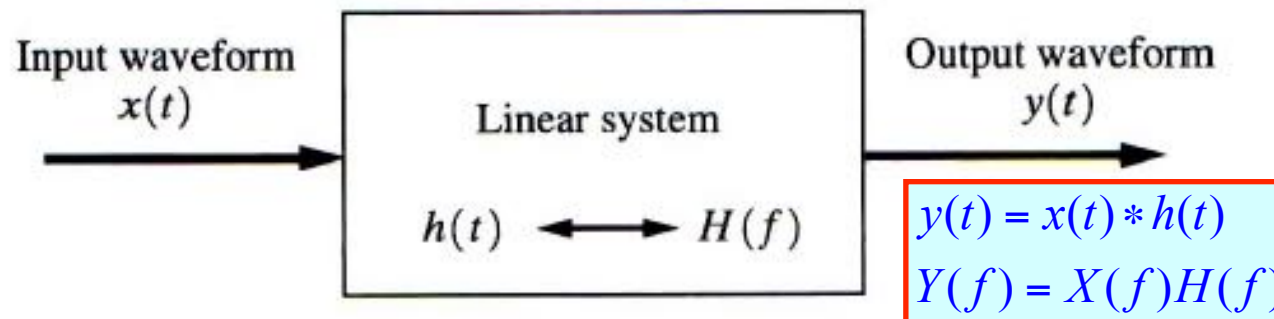
## 2.6 REVIEW OF LINEAR SYSTEMS

### Linear Time-Invariant System

An electronic filter or system is *Linear* when *Superposition* holds, that is when,

$$y(t) = L[a_1x_1(t) + a_2x_2(t)] = a_1L[x_1(t)] + a_2L[x_2(t)]$$

- Where  $y(t)$  is the output and  $x(t) = a_1x_1(t) + a_2x_2(t)$  is the input.
- $L[.]$  denotes the linear system operator acting on  $[.]$ .



## 2.6 REVIEW OF LINEAR SYSTEMS

### Linear Time-Invariant System

Example. Are the following system is linear system or not?

1.)  $y(t) = 5x(t)$

2.)  $y(t) = 5x(t) + 3$

## 2.6 REVIEW OF LINEAR SYSTEMS

### Linear Time-Invariant System

#### Conditions for time-invariance

$Sys\{x(t)\} = y(t)$  implies that  $Sys\{x(t-\tau)\} = y(t-\tau)$

- If the system is time invariant for any delayed input  $x(t - t_0)$ , the output is delayed by just the same amount  $y(t - t_0)$ .
- That is, the shape of the response is the same no matter when the input is applied to the system.

## 2.6 REVIEW OF LINEAR SYSTEMS

### Linear Time-Invariant System

Example. Are the following system is time-invariant system or not?

1.)  $y(t) = 5x(t)$

2.)  $y(t) = 5x(t) + 3$

## 2.6 REVIEW OF LINEAR SYSTEMS

### Impulse Response

The impulse response is the solution to the differential equation when the forcing function is a Dirac delta function. That is

$$y(t) = h(t) \text{ when } x(t) = \delta(t).$$

A general waveform at the input may be approximated by

$$x(t) = \sum_{n=0}^{\infty} x(n\Delta t) [\delta(t - n\Delta t)] \Delta t$$

The output may be approximated by

$$y(t) = \sum_{n=0}^{\infty} x(n\Delta t) [h(t - n\Delta t)] \Delta t$$

This expression becomes the exact result as  $\Delta t$  becomes zero, letting  $n\Delta t = \lambda$ , we obtain

**Convolution** 
$$y(t) = \int_{-\infty}^{\infty} x(\lambda) h(t - \lambda) d\lambda = x(t) * h(t)$$

## 2.6 REVIEW OF LINEAR SYSTEMS

### Transfer Function

- ✧ The output waveform for a time-invariant network can be obtained by convolving the input waveform with the impulse response of the system.
- ✧ The spectrum of the output signal is obtained by taking the Fourier transform of both sides. Using the convolution theorem, we get

$$Y(f) = X(f)H(f)$$

or

$$H(f) = \frac{Y(f)}{X(f)}$$

or  $H(f) = \mathfrak{F}[h(t)]$  is said to be the **transfer function** or **frequency response** of the network.

## 2.6 REVIEW OF LINEAR SYSTEMS

### Transfer Function

The impulse response and frequency response are a Fourier transform pair

$$h(t) \longleftrightarrow H(f)$$

Of course, the transfer function  $H(f)$  is, in general, a complex quantity and can be written in polar form as

$$H(f) = |H(f)| e^{j\angle H(f)}$$

where  $|H(f)|$  is the **amplitude** (or **magnitude**) response and

$$\theta(f) = \angle H(f) = \tan^{-1} \left[ \frac{\text{Im}\{H(f)\}}{\text{Re}\{H(f)\}} \right]$$

is the **phase** response of the network.

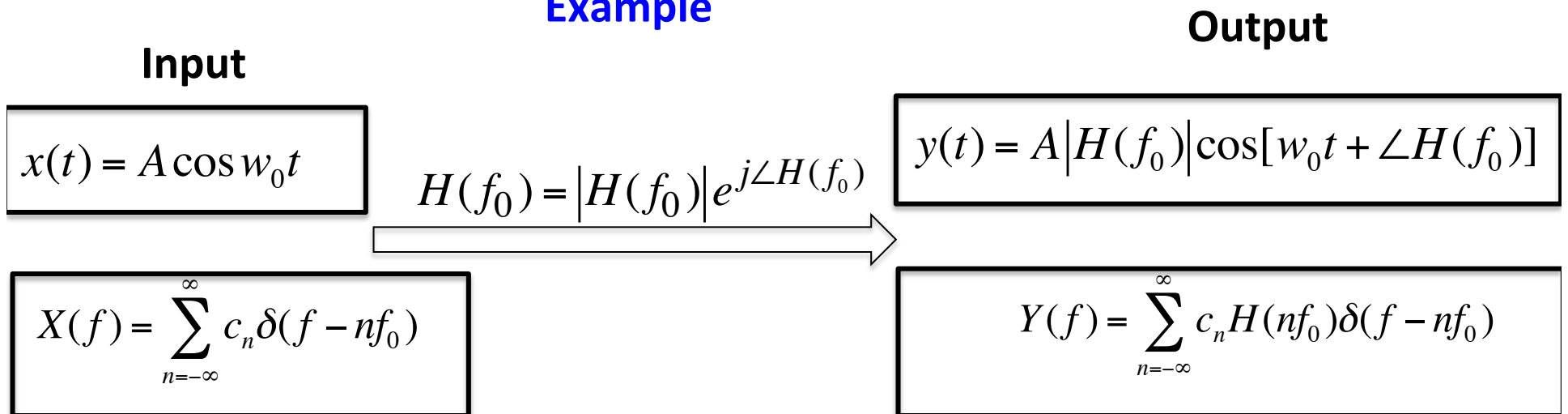
## 2.6 REVIEW OF LINEAR SYSTEMS

### Transfer Function

Since  $h(t)$  is a real function of time (for real networks), it follows

- ✧  $|H(f)|$  is an even function of frequency (  $|H(-f)| = |H(f)|$  )
- ✧  $\vartheta(f)$  is an odd function of frequency. (  $\vartheta(-f) = -\vartheta(f)$  )

### Example





## 2.6 REVIEW OF LINEAR SYSTEMS

### Power Transfer Function

We also can obtain the relationship between the *power spectral density* (**PSD**) at the input, and the output for a linear time-invariant network as:

$$P_y(f) = \lim_{T \rightarrow \infty} \frac{1}{T} |Y_T(f)|^2$$

using

$$Y(f) = X(f)H(f)$$

we get

$$P_y(f) = |H(f)|^2 \lim_{T \rightarrow \infty} \frac{1}{T} |X_T(f)|^2$$

or

$$P_y(f) = |H(f)|^2 P_x(f)$$

**Power transfer function**

$$G_h(f) = \frac{P_y(f)}{P_x(f)} = |H(f)|^2$$

## 2.6 REVIEW OF LINEAR SYSTEMS

### Distortionless Transmission

In communication systems, a distortionless channel is often desired. This implies that the channel output is just proportional to a delayed version of the input

$$y(t) = A(x - T_d)$$

Where  $A$  is the **gain** (which may be less than unity) and  $T_d$  is the **delay**

**In the frequency domain**

$$Y(f) = AX(f)e^{-j2\pi fT_d}$$

**The transfer function of the channel is**

$$H(f) = \frac{Y(f)}{X(f)} = Ae^{-j2\pi fT_d}$$

## 2.6 REVIEW OF LINEAR SYSTEMS

### Distortionless Transmission

For a linear time-invariant system, two requirements are needed as Distortionless system.

✧ The *amplitude response* is flat. That is,

$$|H(f)| = \text{constant} = A$$

### No Amplitude Distortion

✧ The *phase response* is a linear function of frequency. That is

$$\theta(f) = \angle H(f) = -2\pi f T_d$$

### No Phase Distortion

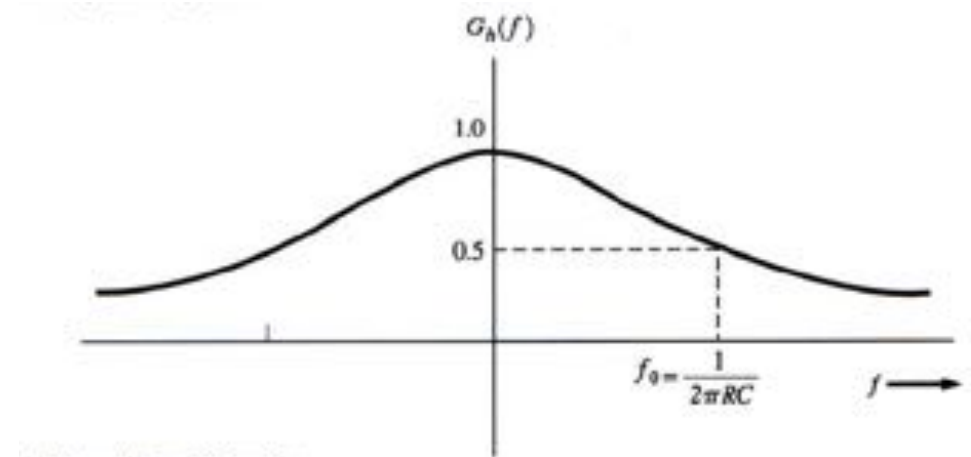
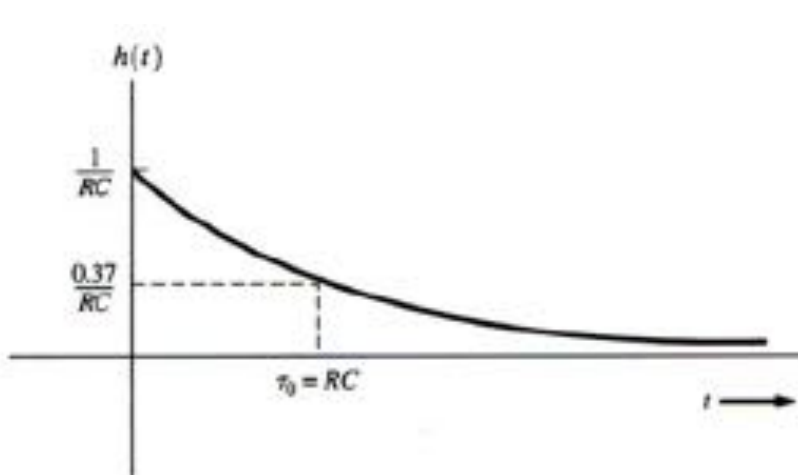
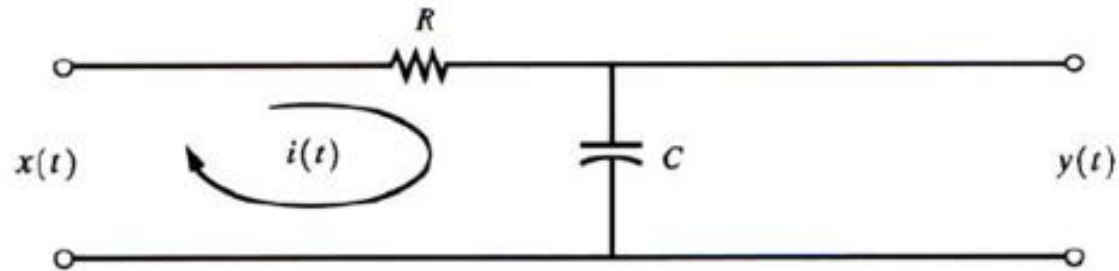
We define the *time delay* of the system as

$$T_d(f) = -\frac{1}{2\pi f} \theta(f) = -\frac{1}{2\pi f} \angle H(f)$$

For distortionless system,  $T_d(f)$  must be **constant**.

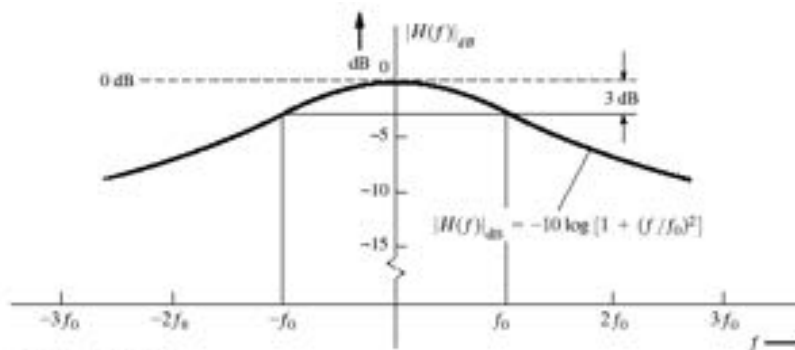
## 2.6 REVIEW OF LINEAR SYSTEMS

### Example 2-18 RC Low-Pass Filter

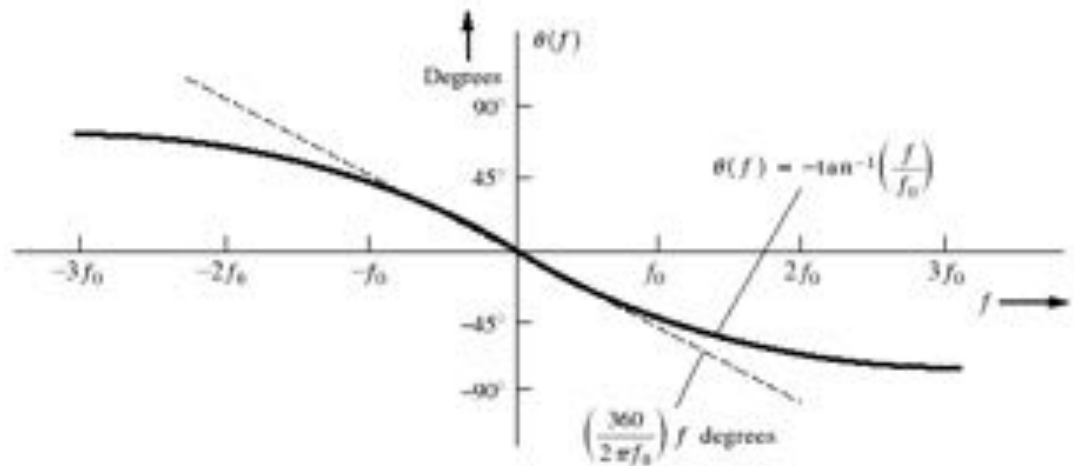


## 2.6 REVIEW OF LINEAR SYSTEMS

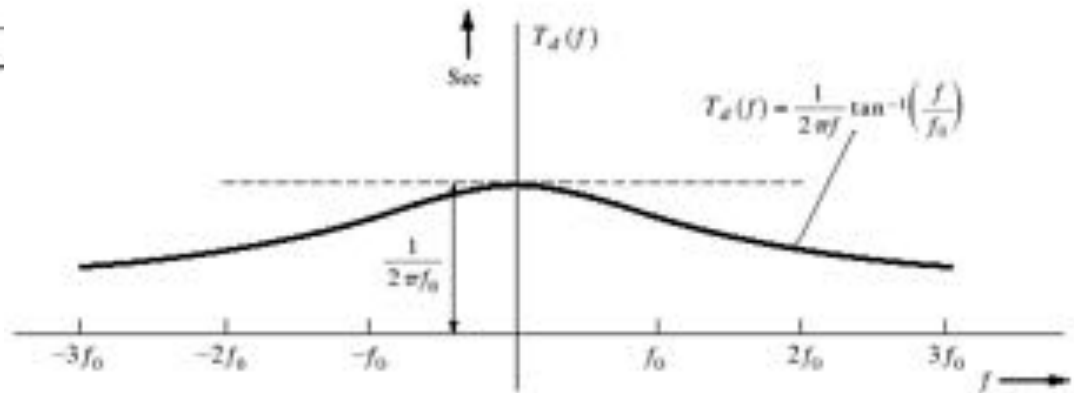
### Example 2-18 Distortion caused by a filter



(a) Magnitude Response



(b) Phase Response



(c) Time Delay

## 2.6 REVIEW OF LINEAR SYSTEMS

### Effects of Distortion

#### Audio

Human ear more sensitive to amplitude distortion, but less so to phase distortion

#### Analog Video

Human visual system more sensitive to time delay errors, which result in smearing of edges, but less so to intensity variation

#### Digital Signals

Pulses smearing into other time slots – “Inter-Symbol Interference” (ISI)